

Technical Comments

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Comment on "Model-Following System with Assignable Error Dynamics and its Application to Aircraft"

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KAWAHATA considers in Ref. 1 the model following problem: given a completely controllable and observable linear time-invariant plant,

$$\begin{bmatrix} \dot{x} \\ y \end{bmatrix} = \begin{bmatrix} F & G \\ H & D \end{bmatrix} \begin{bmatrix} x \\ u \end{bmatrix} \quad (1)$$

and model,

$$\begin{bmatrix} \dot{x}_m \\ y_m \end{bmatrix} = \begin{bmatrix} F_m & G_m \\ H_m & D_m \end{bmatrix} \begin{bmatrix} x_m \\ u_m \end{bmatrix} \quad (2)$$

find the control u which causes y to track y_m for an unspecified u_m . The number of outputs, y , (an $l \times 1$ vector) is assumed to equal the number of controls, u , (an $m \times 1$ vector). Kawahata required that D and D_m in Eqs. (1) and (2) be zero and restricted the control u to have the following form

$$u = K_x x + K_{x_m} x_m + K_{u_m} u_m \quad (3)$$

Additional constraints are imposed in Ref. 1 to generate a solution to the model following problem. K_x , in Eq. (2), assigns some of the closed-loop eigenvalues of the plant to the finite transmission zeroes of the plant, the finite transmission zeroes of the plant must be stable, a special matrix B^* (B^* is defined in Eq. 21 of Ref. 1) must be invertible and $\sigma_k \leq \sigma_{mk}$ for $k=1,2,\dots,l$ (σ_k and σ_{mk} are defined in Eqs. 5a and 6 of Ref. 1). In Kawahata's conclusion, it is indicated that a more general form of Eq. (3) will be investigated in future work where special compensators and derivatives of u_m will be included in Eq. (3).

The purpose of this correspondence is to point out that recently a more general solution to the model following problem has been solved in Refs. 2 and 3. We will briefly show that the model following result in Ref. 1 is a special case of the more general solution. In Ref. 2 it is not required that D and D_m be zero. The more general class of controllers allowed in Ref. 2 includes compensators and derivatives of u_m if they are needed. If simple state feedback is used as the regulation mechanism in the tracking control system it is not required that K_x assign some of the closed-loop eigenvalues to the finite transmission zeroes of the plant. It is possible to do so if desired, however. It is not required in Ref. 2 that the plant

transmission zeroes be stable, provided the unstable transmission zeroes are common to the plant and model (Ref. 4). No relationship is required between σ_k and σ_{mk} , however, derivatives of u_m may become necessary in the tracking control law if $\sigma_k > \sigma_{mk}$. The solution in Ref. 2 is straightforward and is given in state space form. The solution discussed in Ref. 2 and more extensively detailed in Ref. 3 shows that regulation and tracking are two distinct problems and can be solved separately if desired. A summary of the results in Refs. 2 and 3 follows.

Perfect tracking of y_m by y occurs along the trajectories x^* and u^* which satisfy the following equation:

$$\begin{bmatrix} x^* \\ u^* \end{bmatrix} = \begin{bmatrix} S_{11} & S_{12} \\ S_{21} & S_{22} \end{bmatrix} \begin{bmatrix} x_m \\ u_m \end{bmatrix} + \begin{bmatrix} \Omega_{11} \\ \Omega_{21} \end{bmatrix} v \quad (4)$$

$$\Omega_{11} \dot{v} = v - S_{12} \dot{u}_m \quad (5)$$

The feedforward matrices, S_{11} , S_{12} , S_{21} , and S_{22} satisfy Eq. (9) in Ref. 2, an algebraic equation. A solution to Eq. (9) in Ref. 2 requires that no transmission zero of the plant equal an eigenvalue of the model (see also Ref. 5). Kawahata does not discuss this requirement because the feedback gain in Ref. 1 performs a pole-transmission zero cancellation before solving the model-following problem. The other matrices in Eqs. (4) and (5), Ω_{11} and Ω_{21} , satisfy Eq. (14) in Ref. 2.

The x^* and u^* trajectories, by construction, satisfy the plant dynamics. If the error quantities, $\tilde{x} = (x - x^*)$, and $\tilde{u} = (u - u^*)$, are formed, as discussed in Ref. 3, they satisfy the equation

$$\dot{\tilde{x}} = F\tilde{x} + G\tilde{u} \quad (6)$$

i.e., the plant dynamics. At this point, any control law employing some form of state feedback to stabilize the tracking error can be designed for \tilde{u} that a designer wishes. The particular type of feedback used in the model following control law derived by Kawahata is not explicitly discussed in Refs. 2 and 3 since the objective of Ref. 2 is to solve the model following problem. Model following and feedback control are distinct problems. Feedback control changes \tilde{x} but does not alter x^* .

In order to obtain the control law form chosen by Kawahata [Eq. (3)], simple state feedback is combined with the model following solution

$$\tilde{u} = K_x \tilde{x} \quad (7)$$

$$\dot{\tilde{x}} = (F + GK_x) \tilde{x} \quad (8)$$

Substituting Eq. (4) into Eq. (7) and combining terms produces

$$u = K_x x + (S_{21} - K_x S_{11}) x_m + (S_{22} - K_x S_{12}) u_m + (\Omega_{21} - K_x \Omega_{11}) v \quad (9)$$

K_x in Eq. (9) can be chosen to place closed-loop eigenvalues of the plant to any desired set of symmetric stable eigenvalues. A linear quadratic regulator (LQR) can be used to obtain K_x , if desired, by weighting \tilde{x} and \tilde{u} in the LQR cost function (Ref. 3). Perfect tracking occurs asymptotically (i.e., $\tilde{x} \rightarrow 0$) if the plant and model initial conditions are arbitrary.

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Table 1 Inverse and feedforward matrix solutions

$S_{11} =$	$\begin{bmatrix} 1.0 & 0.0 & 0.0 & 0.0 & 0.0 \\ -0.119 & 0.297 & -1.37 \times 10^{-8} & 0.812 & 0.0 \\ 0.00263 & 0.00444 & 0.388 & 0.0217 & -8.31 \times 10^{-8} \\ -0.00221 & -0.00931 & 0.0 & 1.02 & 0.0 \\ 13.4 & -23.5 & -3.75 \times 10^{-7} & 3.82 & 0.0128 \end{bmatrix}$
$S_{12} = 0$	
$S_{21} =$	$\begin{bmatrix} 0.00126 & -0.00371 & -0.103 & -0.0112 & 0.0 \\ 0.481 & -0.283 & -39.5 & -3.19 & 1.72 \times 10^{-4} \end{bmatrix}$
$S_{22} =$	$\begin{bmatrix} 0.0186 & 1.062 \times 10^{-5} \\ 3.52 \times 10^{-9} & 0.319 \end{bmatrix}$
$\Omega_{11} =$	$\begin{bmatrix} 0.0 & 0.0 & 0.0 & 0.0 & 0.0 \\ 0.0 & -0.7924712 & 0.0 & 42.476485 & 0.0 \\ 0.0 & 0.0 & 0.0 & 1.0 & 0.0 \\ 0.0 & -0.0147849 & 0.0 & 0.7924712 & 0.0 \\ 340.13605 & -21.552343 & 0.0 & 1155.2063 & 0.0 \end{bmatrix}$
$\Omega_{21} =$	$\begin{bmatrix} 0.0 & 0.00883 & -0.128 & -0.685 & 0.0 \\ 8.50 & -0.539 & 0.0 & 28.9 & 0.025 \end{bmatrix}$

It can be shown, using results from Ref. 6, that Kawahata eliminates the last term in Eq. (9) by choosing some elements in K_x so that

$$(\Omega_{21} - K_x \Omega_{11})v = 0 \quad (10)$$

The minimal realization of the generalized state space compensator in Eq. (5) (the dynamical equation is also sometimes referred to as a descriptor system) cancels transmission zeroes of the plant not common between the plant and model. If some of the elements in K_x can be chosen to satisfy Eq. (10), then some of the closed-loop eigenvalues of the plant equal all of the *finite* plant transmission zeroes as discussed by Kawahata. Plant transmission zero locations do not always satisfy desirable handling qualities specifications.

Reworking the example in Ref. 1 using Eqs. (4) and (9) where the plant is a Beechcraft Model 65 and the model to be followed is that of a Boeing 747 yields the results in Table 1. Using Table 1, and K_x from Table 2 in Ref. 1, the values for $S_{21} - K_x S_{11}$ and $S_{22} - K_x S_{12}$ are in agreement with K_{xm} and K_{um} in Table 2 of Ref. 1 to three significant digits. Note that S_{12} is the null matrix, causing the minimal realization of v to be $v = 0$. The last term in Eq. (9) is zero for any value of K_x , a feature of this example that is also pointed out by Kawahata.

References

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Reply by Author to J.R. Broussard and L.E. Mabius

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I WISH to thank Mr. Broussard and Mr. Mabius for their Comment on my work¹ and for drawing attention to their Ref. 2. When my manuscript was prepared, Ref. 2 was not yet available. Even so, the two papers complement each other.

In Ref. 1, a practically oriented design technique was stressed rather than a more theoretically oriented approach to the model following system. The appearance of time

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